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Novel Speed Sensorless DT/SC-SVM Scheme for Induction Motor Drives

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Presentation outline

- * Introduction
- * Direct Torque/Slip Control Method
- * Field Weakening Operation
- * Flux and speed estimation
- * Experimental Results
- * Final Conclusions





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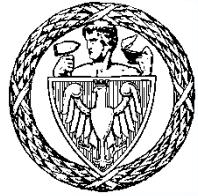


Requirements

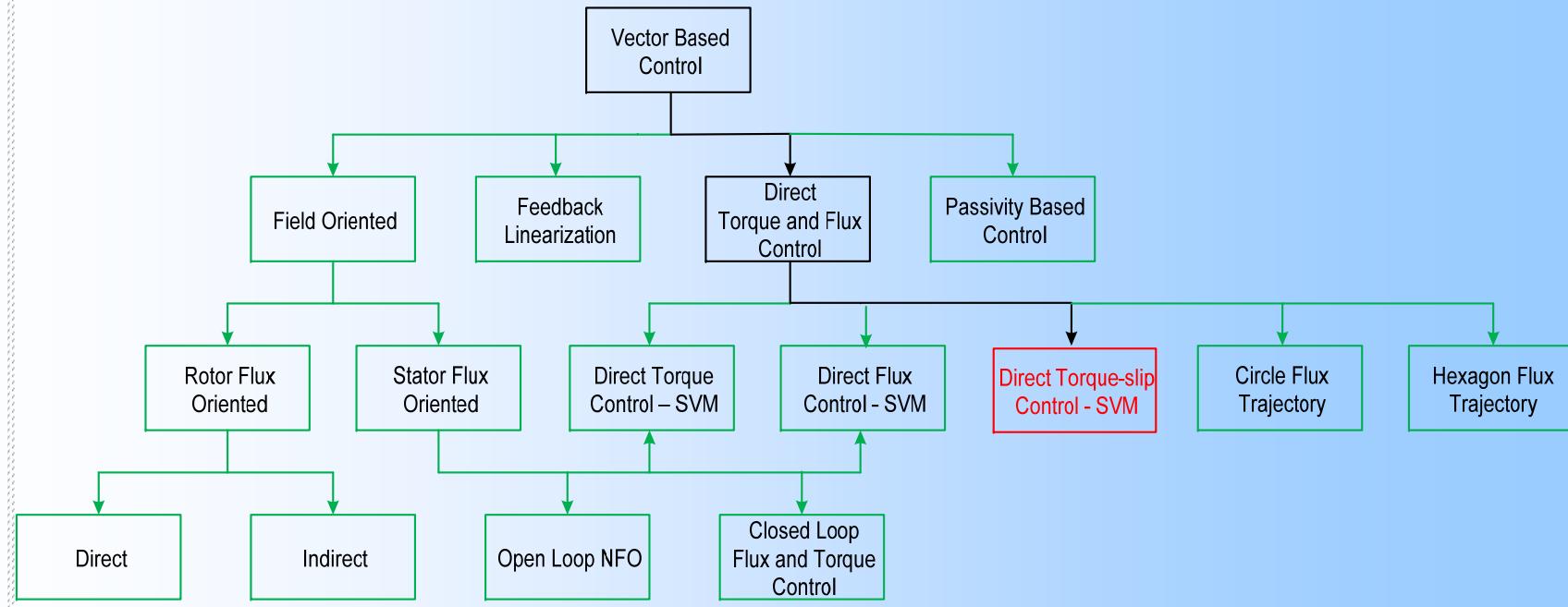
Main conditions what should be met to achieve good IM drive for trams and metro:

- wide range of speed control,
- high dynamic of flux and torque,
- high starting torque,
- maximum use of available torque in flux weakening region,
- low sensitivity to motor parameters changes,
- low torque ripples,
- minimization of switching losses :
 - constant and low switching frequency,
 - unipolar inverter output voltage (elimination of $+/- U_{dc}$) switchings.



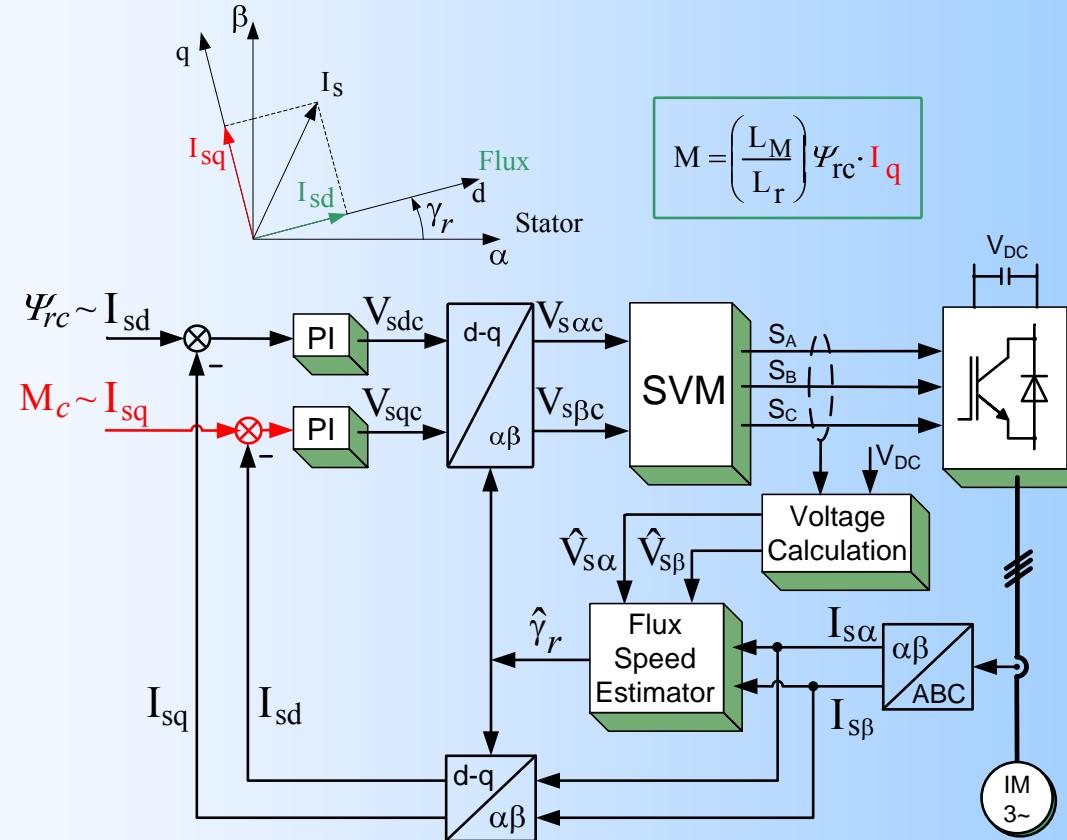


Classification of high performance IM control methods





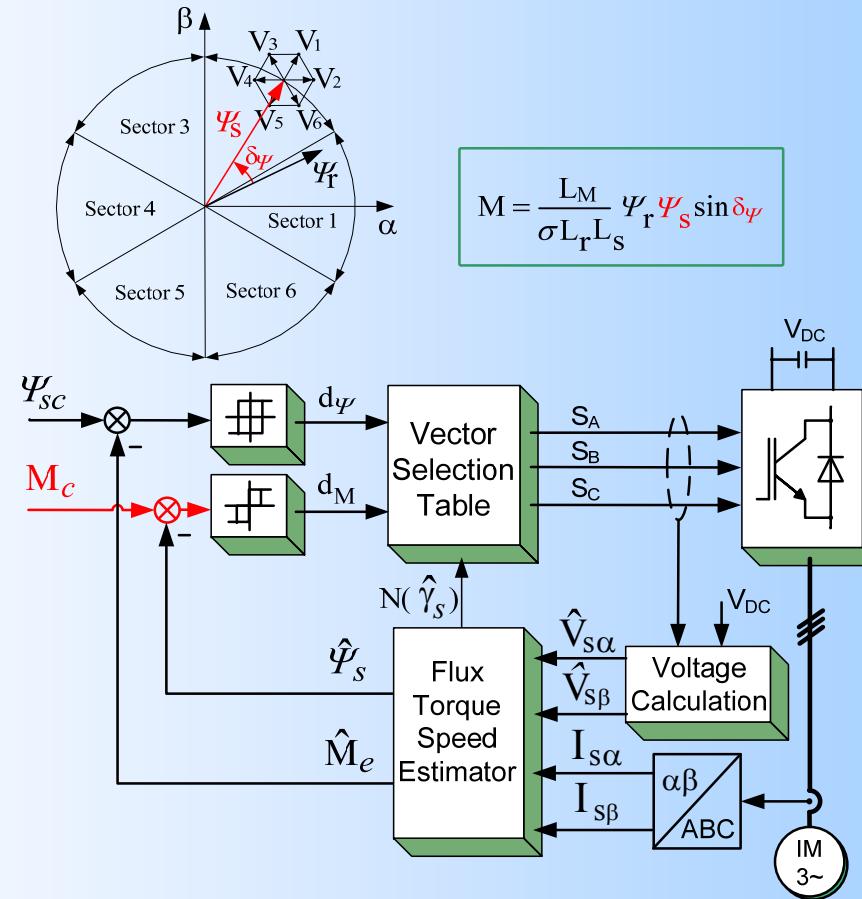
High performance FOC control



Vector diagram and block scheme of rotor FOC. Torque is controlled indirectly via torque current I_{sq} control loop



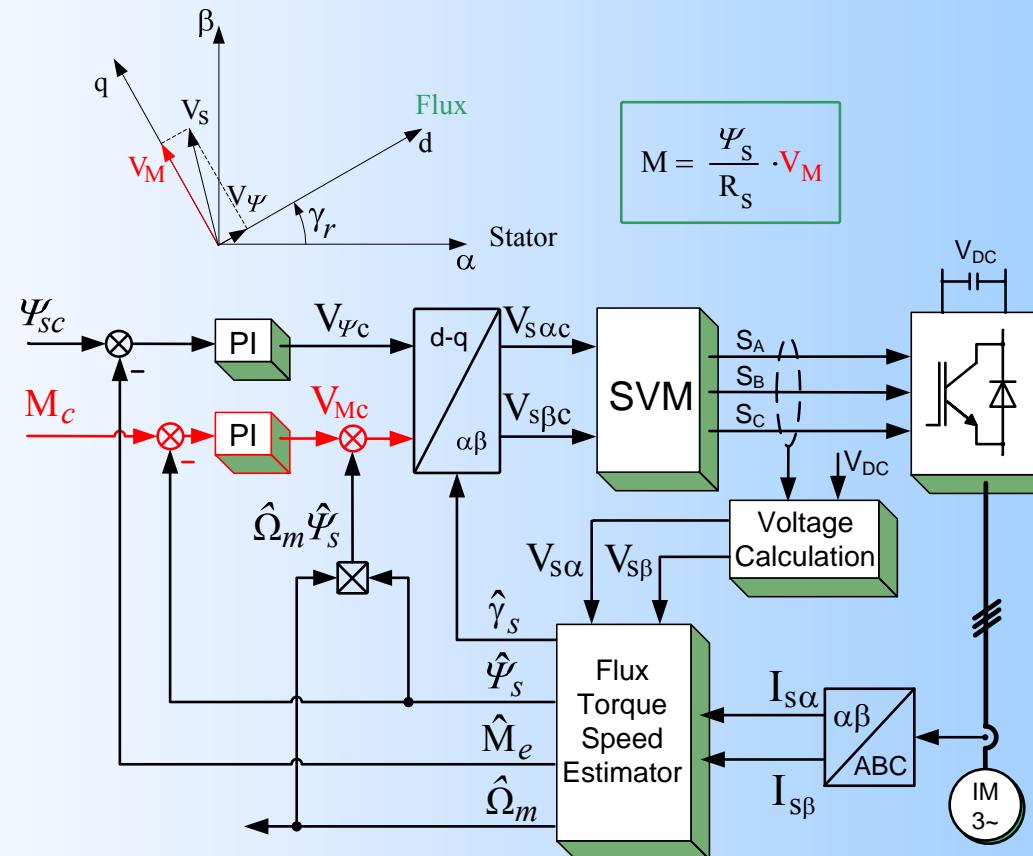
High performance ST-DTC control



In Switching Table based DTC torque is controlled *directly* via stator flux vector movement by selection of appropriate forward/backward active voltage vector (V_1 or V_6) and stops by selection zero voltage vector V_0 . Stator flux vector moves on circular path



High performance DTC-SVM control



Vector diagram and block scheme of DTC-SVM.
Torque is controlled *directly* via voltage vector component V_M



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Direct Torque/Slip Control with Space Vector Modulation (DT/SC-SVM)

$$U_{Sy} = R_S \cdot I_{Sy} + \Omega_S \cdot \Psi_S$$

$$M_e = p_b \cdot \frac{m_s}{2} \cdot \Psi_S \cdot I_{Sy}$$

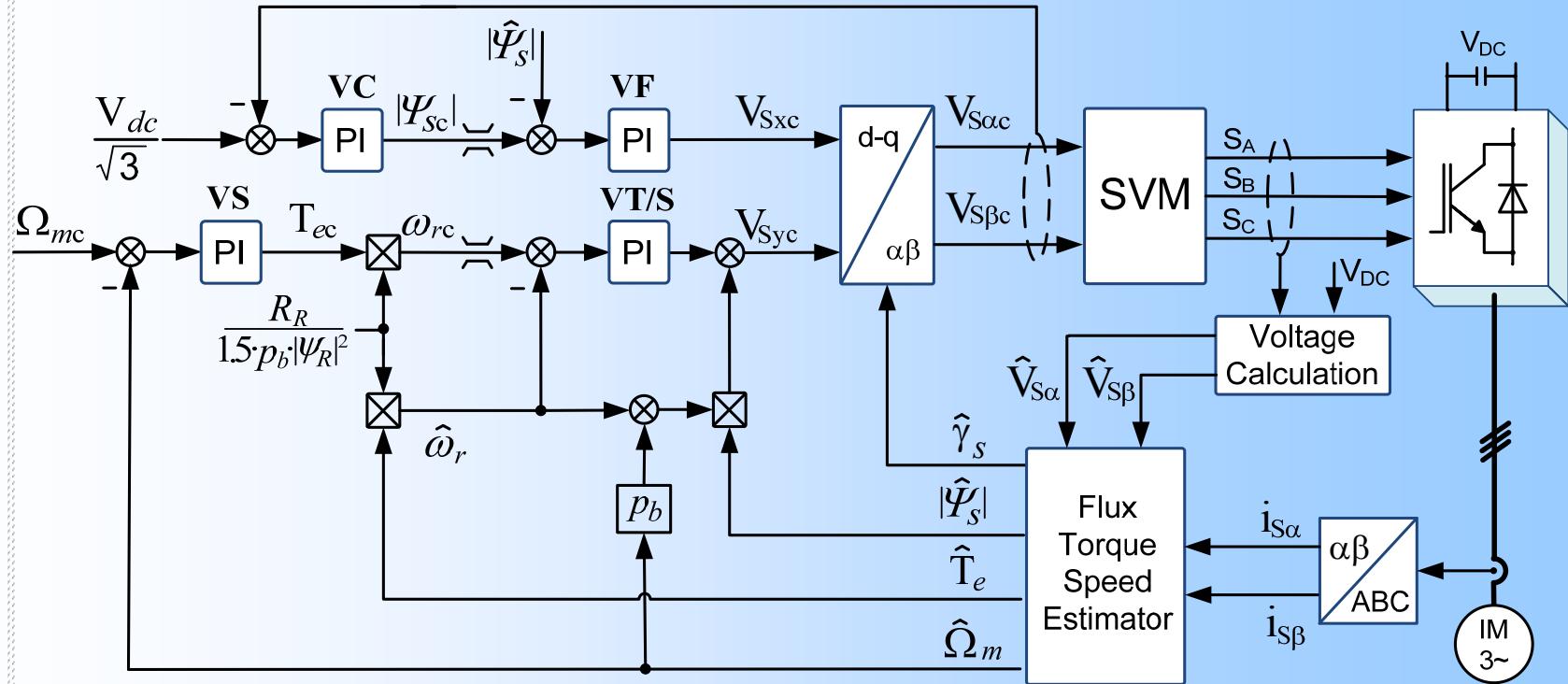
$$\omega_r = \frac{R_R}{\frac{m_s}{2} \cdot p_b \cdot |\Psi_R|^2} \cdot M_e$$

$$U_{Sy} = \frac{R_S}{p_b \cdot \frac{m_s}{2} \cdot \Psi_S} \cdot M_e + (\omega_r + \Omega_m \cdot p_b) \cdot \Psi_S$$





Direct Torque/Slip Control with Space Vector Modulation (DT/SC-SVM)



DT/sC-SVM method with direct torque/slip control and flux weakening. VC – voltage controller, VS – speed controller, VF – flux controller, VT/S – torque/speed controller



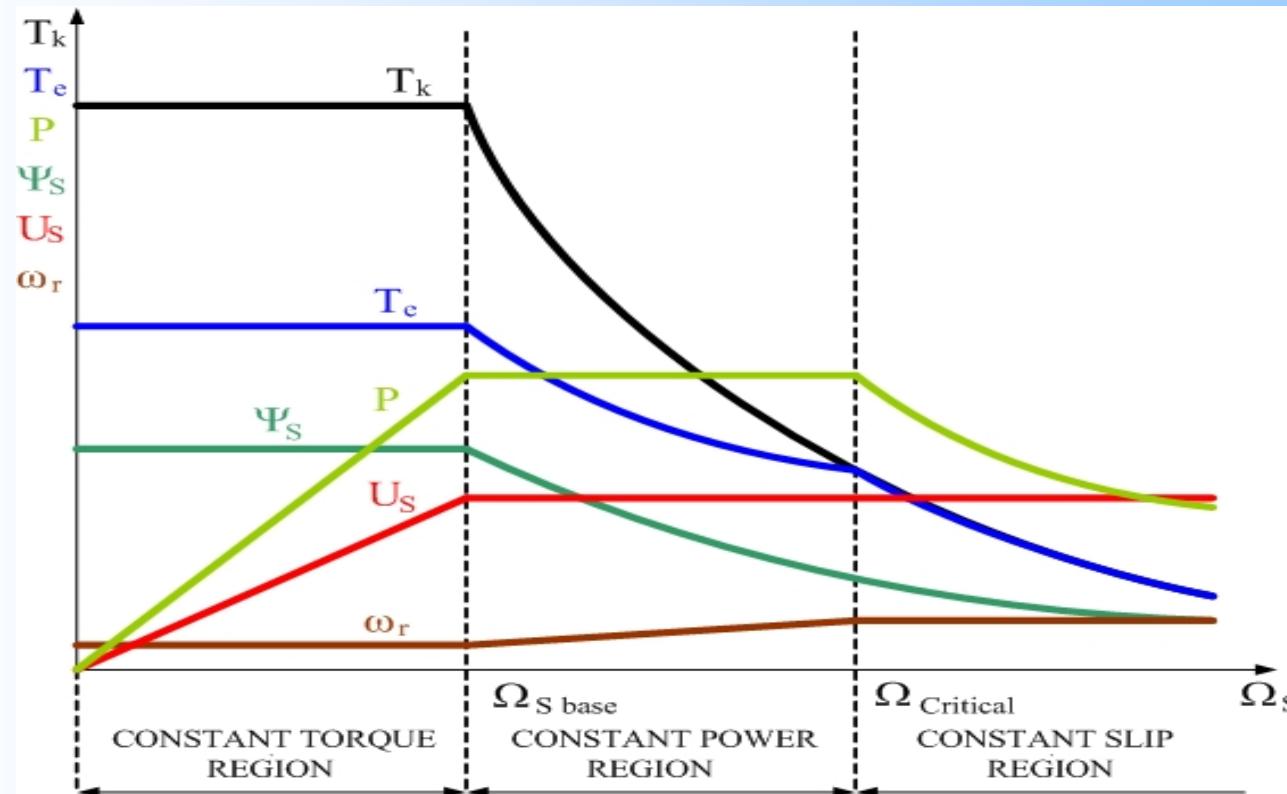
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Field Weakening Operation

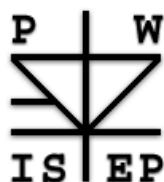


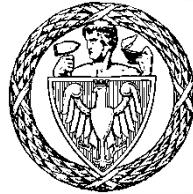
- operation at constant power and constant slip region ,
- maximum use of available DC voltage,
- high dynamic of torque generation .



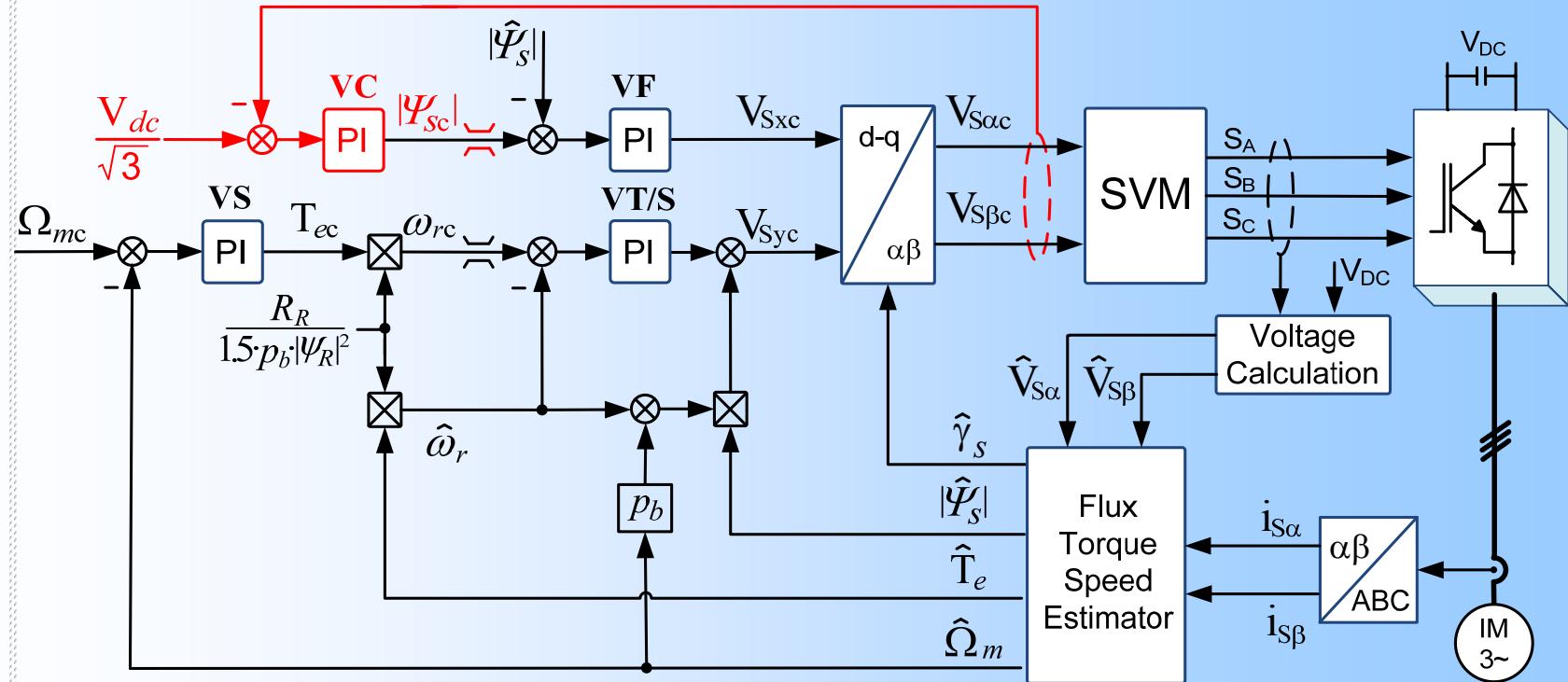
Field Weakening Techniques

- * V/Hz adjustment (Scalar Control)
- * Flux Program Reference $1/\Omega_m$ (Vector Control)
- * Maximum Torque Capability (Vector Control)
- * Operation at Voltage limit
- * **Sensorless Voltage Controller**





Field Weakening in DT/S-SVM Control Method



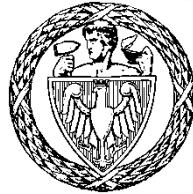
DT/SC-SVM scheme operates in three drive ranges:
constant torque ($M_e = \text{const.}$), constant power ($P = \text{const.}$)
and constant slip ($\omega_r = \text{const.}$)



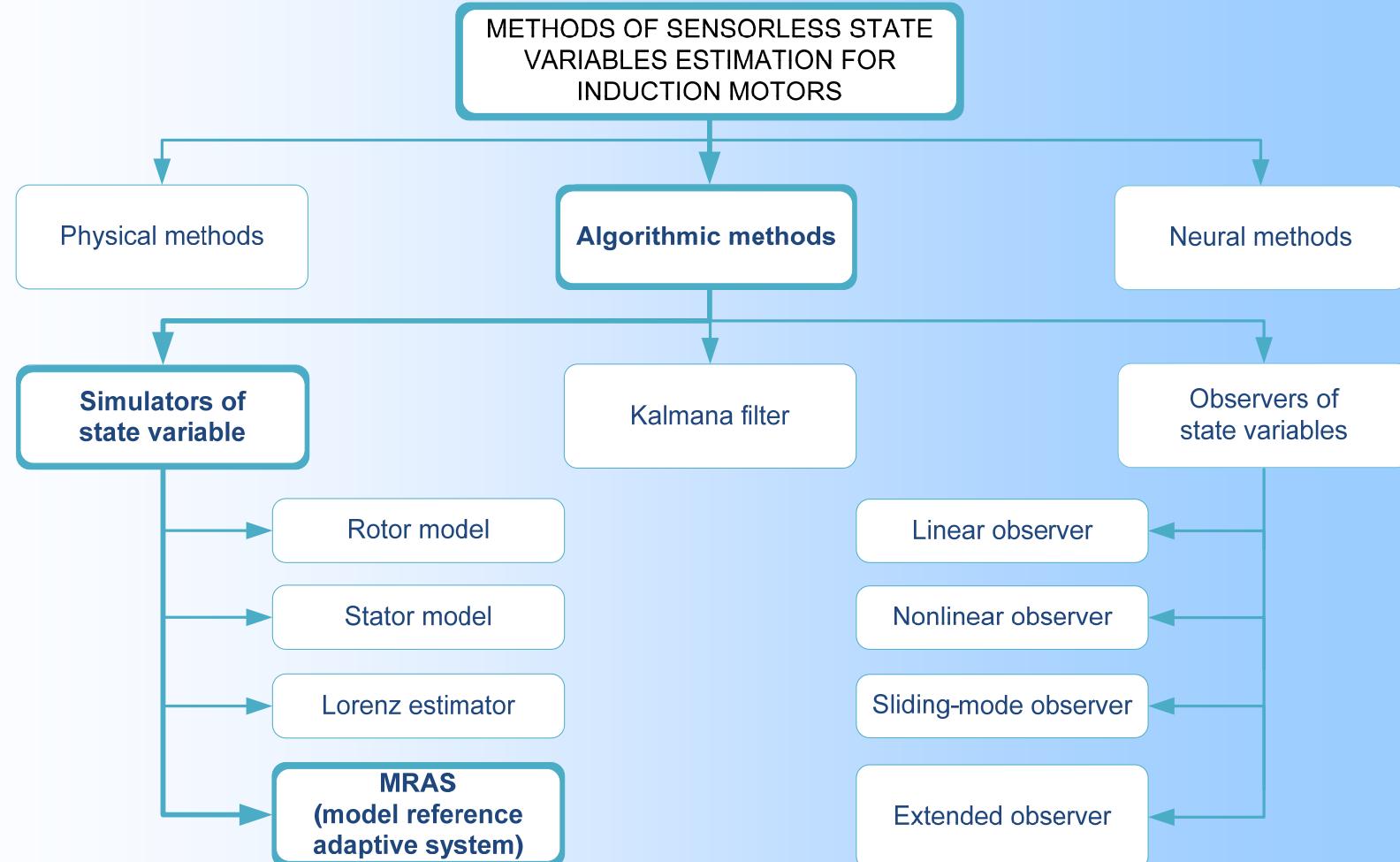
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Classification of flux & speed estimators





Flux and speed estimator

Flux estimation

$$\frac{d\hat{\Psi}_{S\alpha}}{dt} = V_{S\alpha} - \hat{I}_{S\alpha} R_S + e_{I\alpha}$$

$$\frac{d\hat{\Psi}_{S\beta}}{dt} = V_{S\beta} - \hat{I}_{S\beta} R_S + e_{I\beta}$$

$$\frac{d\hat{\Psi}_{R\alpha}}{dt} = \left(\frac{R_S L_S}{L_M}\right) \hat{I}_{S\alpha} - \frac{R_R}{L_M \hat{\Psi}_{S\alpha}} - p_b \Omega_m \hat{\Psi}_{R\beta}$$

$$\frac{d\hat{\Psi}_{R\beta}}{dt} = \left(\frac{R_S L_S}{L_M}\right) \hat{I}_{S\beta} - \frac{R_R}{L_M \hat{\Psi}_{S\beta}} + j p_b \Omega_m \hat{\Psi}_{R\alpha}$$

$$\hat{I}_{S\alpha} = \frac{(L_R \hat{\Psi}_{S\alpha} - L_M \hat{\Psi}_{R\alpha})}{(L_R L_S \sigma)}$$

$$\hat{I}_{S\beta} = \frac{(L_R \hat{\Psi}_{S\beta} - L_M \hat{\Psi}_{R\beta})}{(L_R L_S \sigma)}$$

Speed estimation

$$\Delta I_{S\alpha} = I_{S\alpha} - \hat{I}_{S\alpha}$$

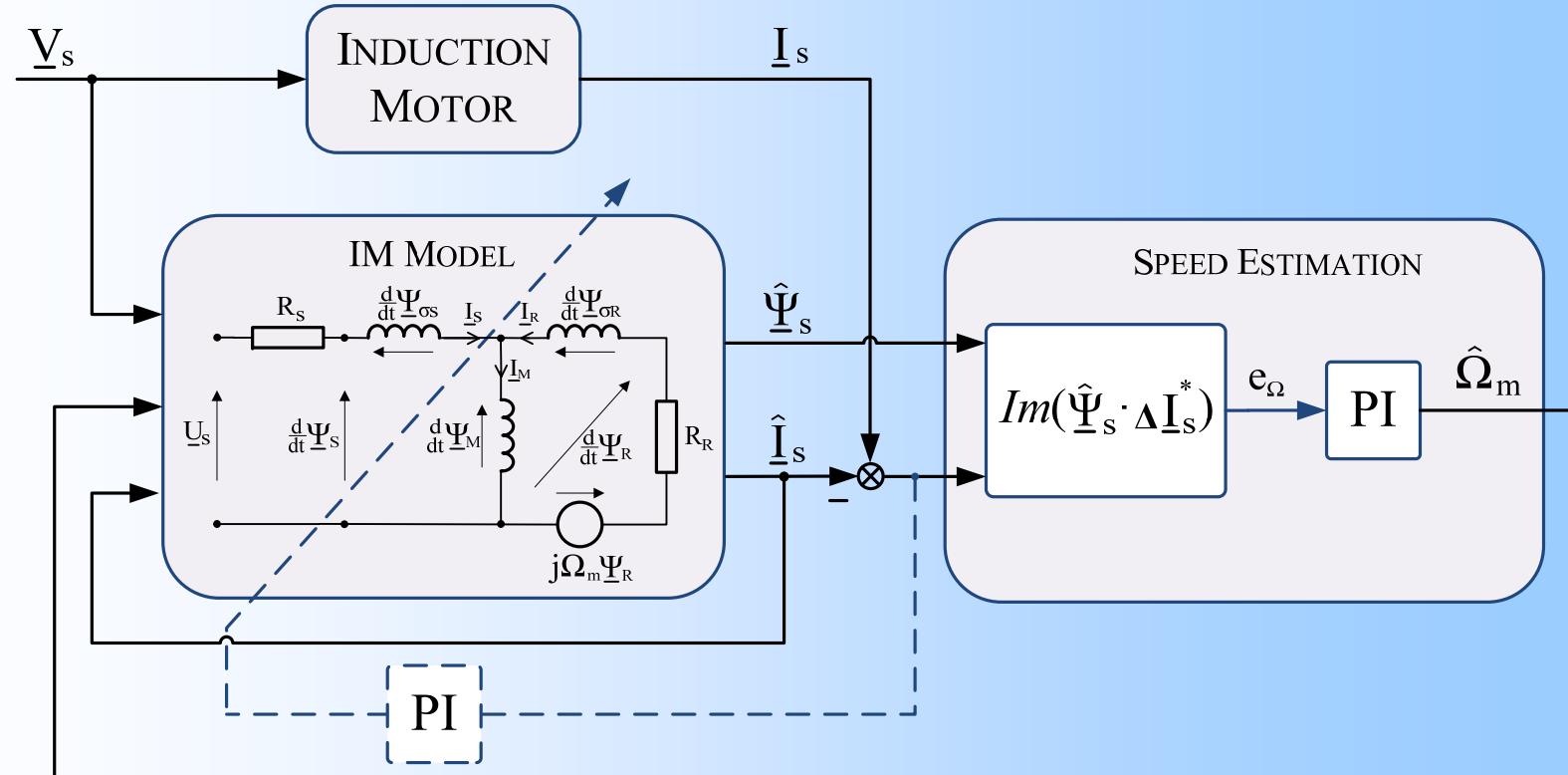
$$\Delta I_{S\beta} = I_{S\beta} - \hat{I}_{S\beta}$$

$$e_\Omega = \hat{\Psi}_{S\beta} \cdot \Delta I_{S\alpha} - \hat{\Psi}_{S\alpha} \cdot \Delta I_{S\beta}$$

$$\Omega_m = K_P e_\Omega + K_P / T_I \int e_\Omega dt$$



Flux and speed estimator





Advantages and disadvantages of proposed solution

Advantages:

- elimination of speed sensor,
- low complexity of flux and speed estimator,
- direct slip control,
- wide range of speed operation,
- low sensitivity of field weakening to motor parameter changes,
- adaptation of field weakening method to available voltage in DC link, speed and load torque,
- high dynamic of flux and torque generation.

Disadvantages:

- higher sensitivity to motor parameters changes than with speed sensor use,
- additional PI controller in field weakening path





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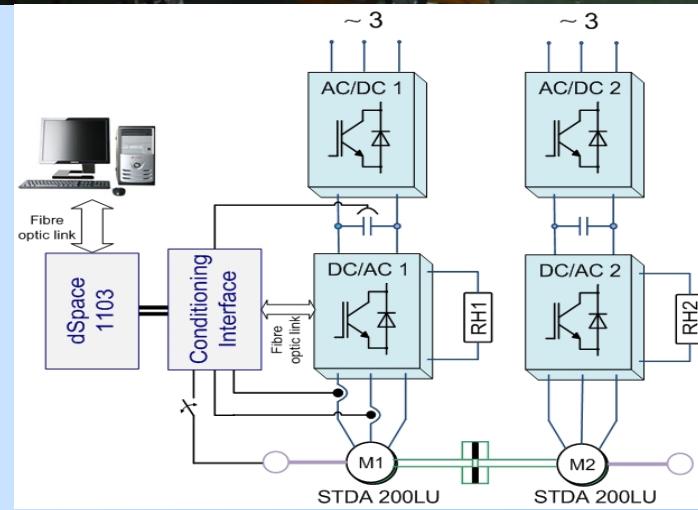




Laboratory setup

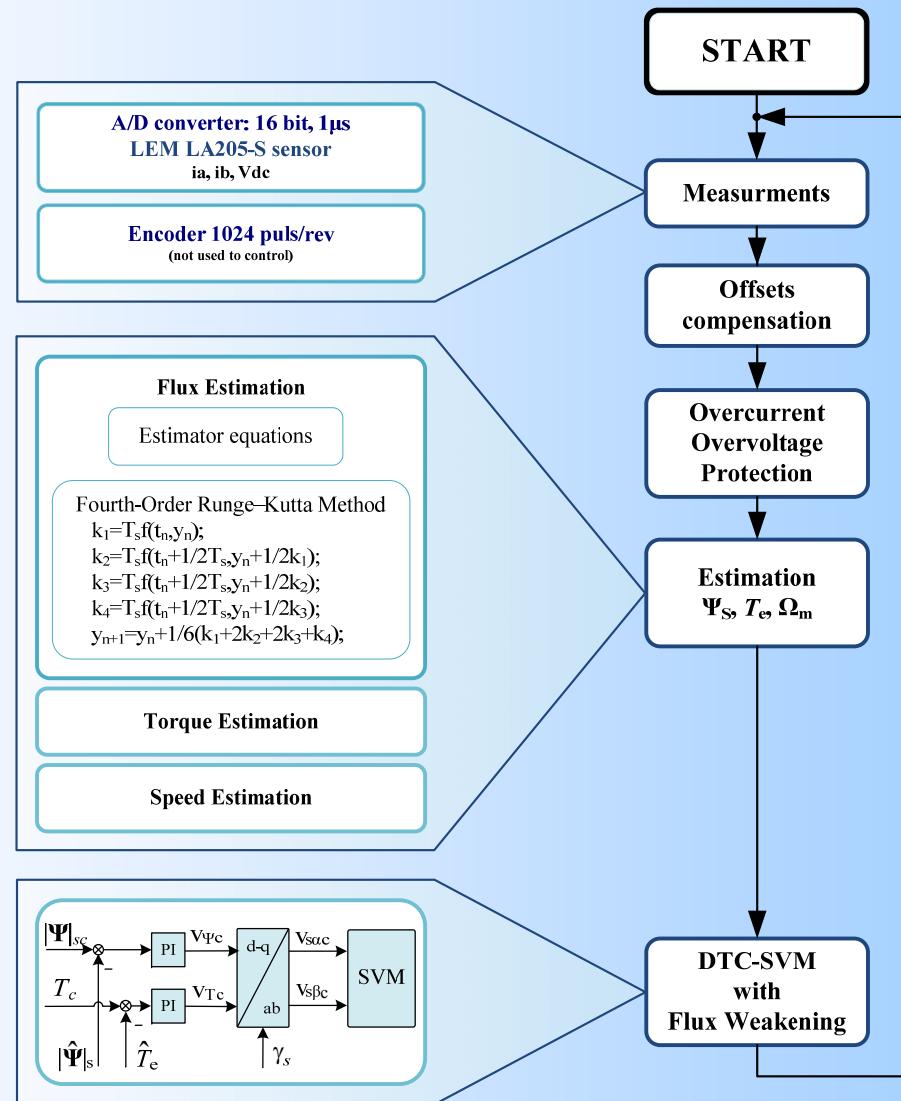
| IM type STDA 200LU | | | |
|--------------------|---------|-------|----------------------|
| P_N | 50 kW | R_S | 64,5 mΩ |
| U_N | 3 x 380 | R_R | 46,3 mΩ |
| I_N | 88 A | L_S | 25,217 mH |
| f_N | 65 Hz | L_R | 25,137 mH |
| M_{eN} | 249 Nm | L_M | 24,75 mH |
| p | 2 | J | 10 kg·m ² |

| Power converter AC/DC and DC/AC | | | |
|---------------------------------|------------|--|--|
| P_N | 55 kW | | |
| I_N | 98 A | | |
| U_N | 3x400 50Hz | | |
| f_{imp} | 4 kHz | | |





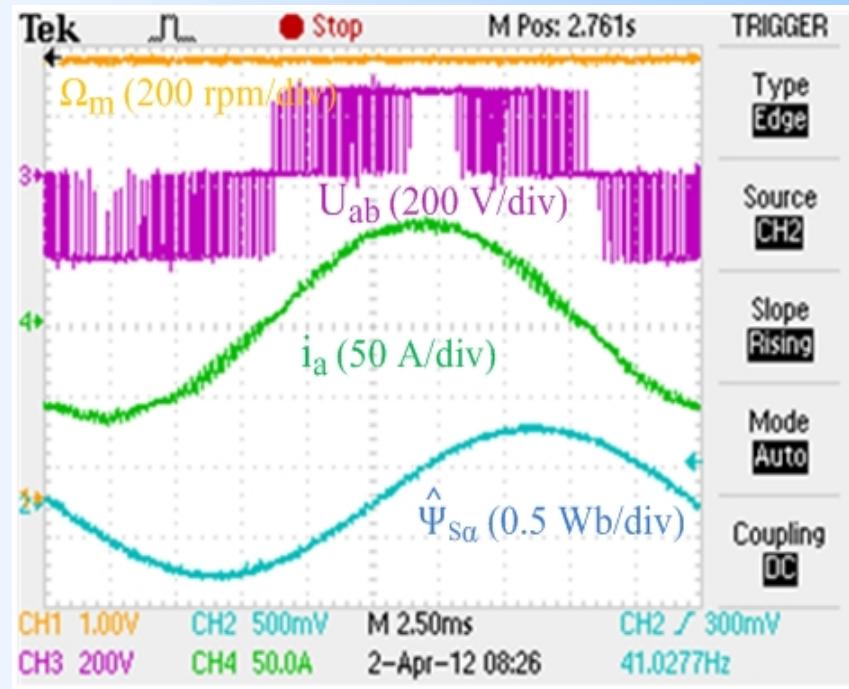
Flow diagram of control algorithm





Experimental Results

Sady state operation



Steady state operation in field weakening
region at 100 Nm load torque



Experimental Results

Steady state operation



Measurement of the stator current THD
GOSEN type Mavowatt 50 M816A
Current probe Metraflex 3003



Experimental Results

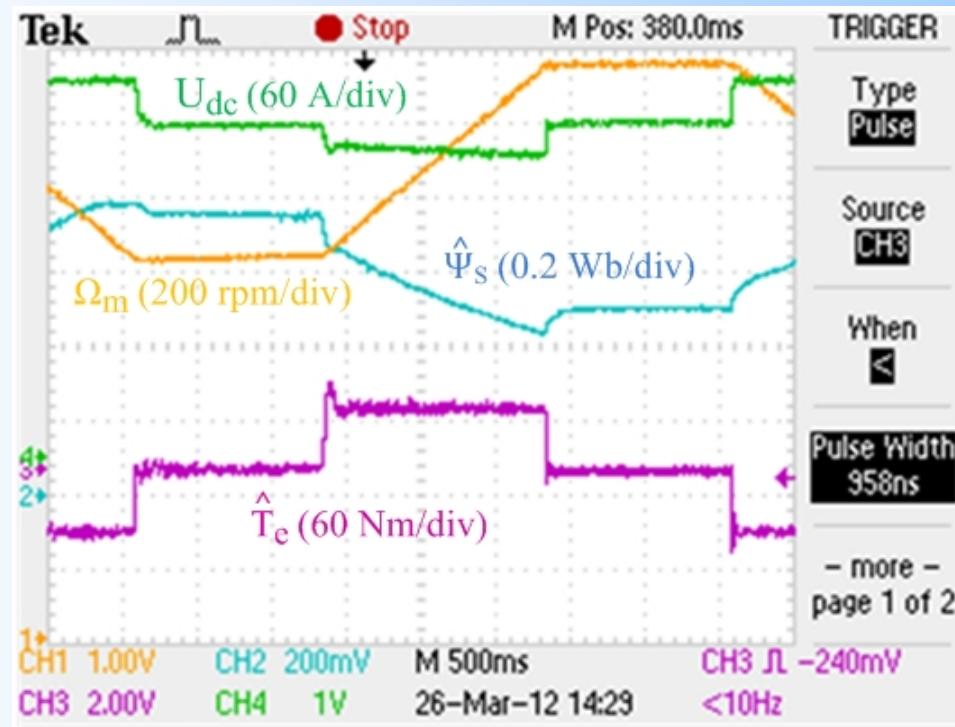
Stator current THD

| Lp. | U_{dc} [V] | f_s [Hz] | I_A [A] | THD [%] | Comments |
|-----|-----------------|---------------|--------------|------------|------------|
| 1 | 625 | 30 | 26 | 5.8 | No load |
| 2 | 625 | 30 | 88 | 1.9 | Rated load |
| 3 | 625 | 50 | 28 | 5.4 | No load |
| 4 | 625 | 50 | 88 | 1,9 | Rated load |

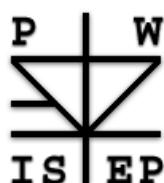


Experimental Results

Constant torque and field weakening region



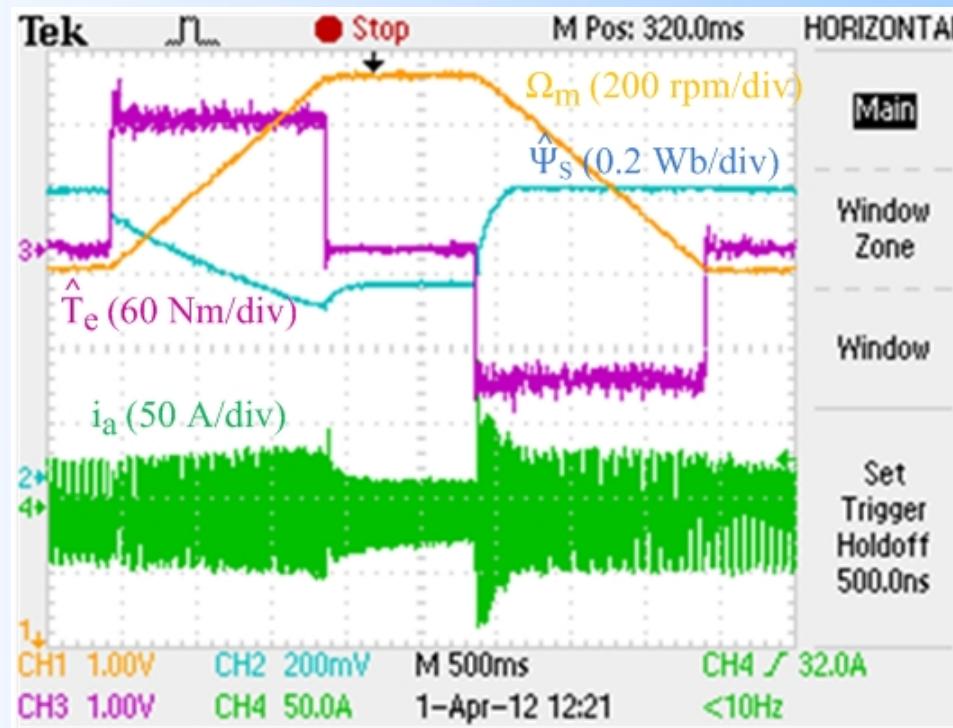
Operation in torque/slip control mode:
torque reference changes +/-50 Nm
Speed changes from 1000 rpm to 1500 rpm





Experimental Results

Constant torque and field weakening region



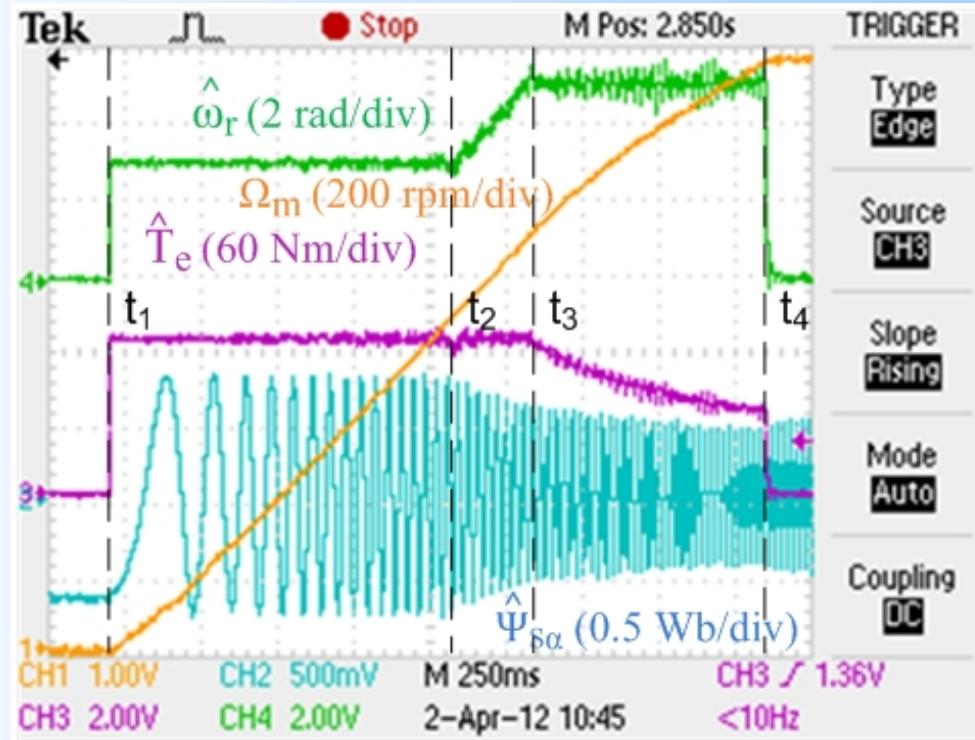
Operation in torque/slip control mode:
torque reference changes +/-100 Nm
Speed changes from 1000 rpm to 1500 rpm





Experimental Results

Constant torque and field weakening region

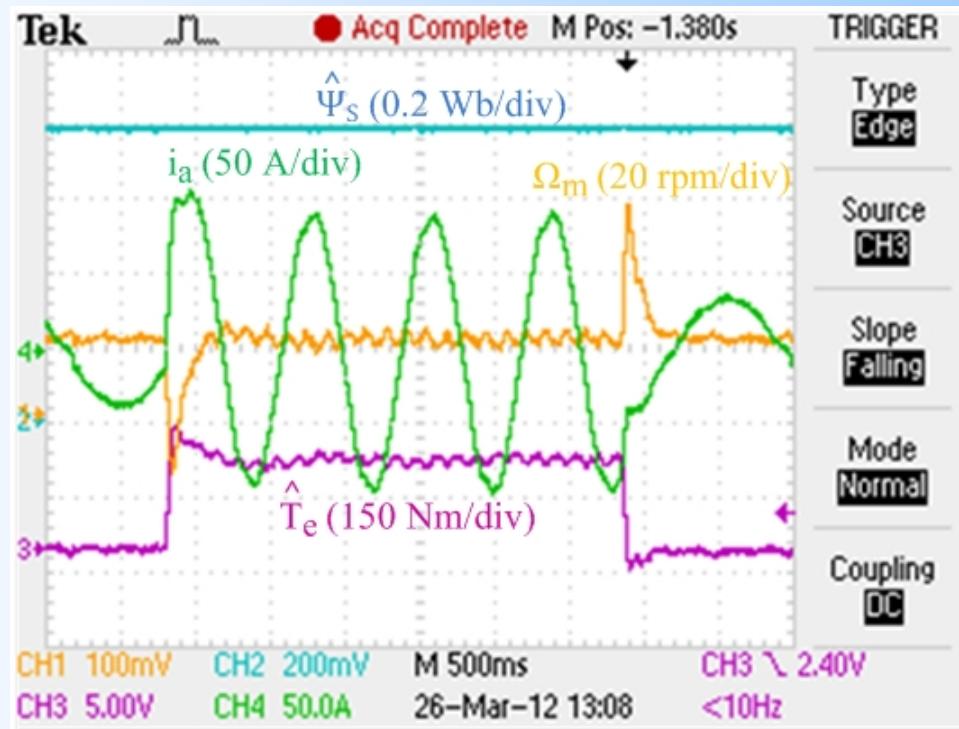


Start up from 0 up to 1580 rpm



Experimental Results

Speed response

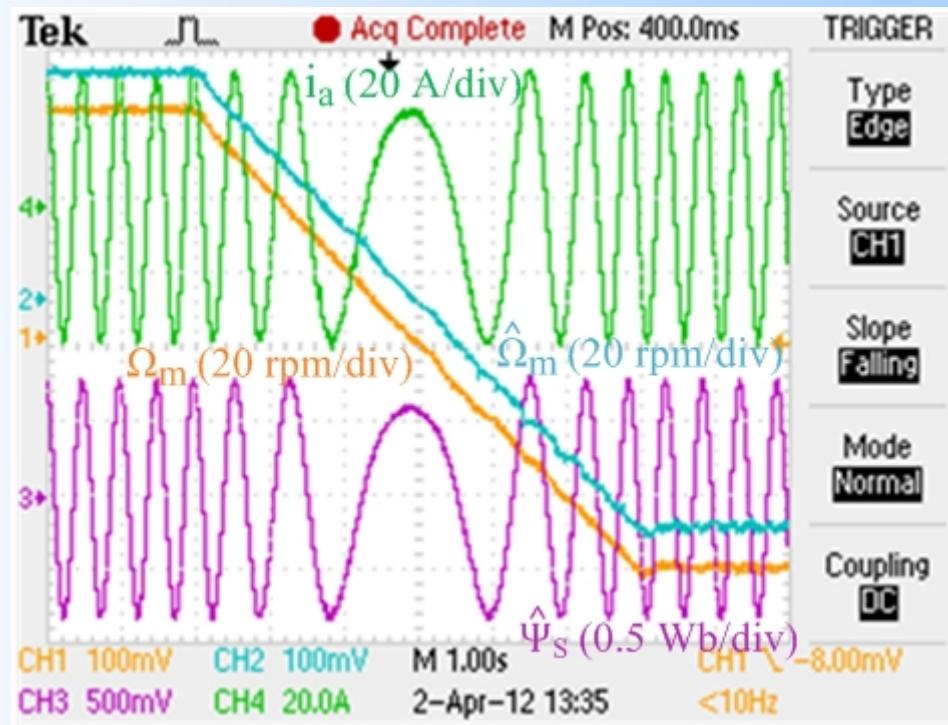


Speed response to the load step change
in speed control mode
Speed 20 rpm/min, Load torque 120 Nm



Experimental Results

Speed reversal



Speed tracking performance for
slow reference changes +/-60 rpm
in the time period of 6s





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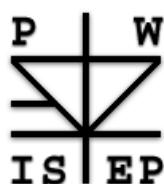
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Final Conclusions

- ⇒ Presented control method allows for direct slip control
- ⇒ Simple flux weakening method allowing for adaptation to all conditions of speed, load torque and available voltage in DC link
- ⇒ Maximum use of available DC voltage
- ⇒ High dynamic of torque generation
- ⇒ Proposed simple flux and speed estimator operates in wide range of speed
- ⇒ Elimination of speed sensor
- ⇒ Whole control system was tested on 50 kW tram IM motor



JOURNAL PAPERS

- P. Wójcik, M.P. Kazmierkowski: „Simple Direct Flux Vector Control with Space Vector Modulation for PWM Inverter Fed Induction Motor Drive”, *Przegląd Elektrotechniczny (Electrotechnical Review)* ISSN 0033-2097, R. 86, Nr 2, **2010**, pp. 60-64 (in English).
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- H. Abu-Rub, D. Stando and M. P. Kazmierkowski, „Simple speed sensorless DTC-SVM scheme for induction motor drives”, *Bulletin of the Polish Academy of Sciences: Technical Sciences*, vol. 60, no. 2, **2013**, pp. 301-307 (in English).

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- D. Stando, M. P. Kazmierkowski, “Novel Speed Sensorless DTC-SVM Scheme for Induction Motor Drives”, in *Proc. 8th International Conference-Workshop Compatibility in Power Electronics, CPE’13*, Ljubljana, June 3-5, Slovenia, **2013** (on ieeexplore)



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